

Lecture notes of Image Compression and Video  
Compression series [2005]

### 3. Subband Coding

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# Topics

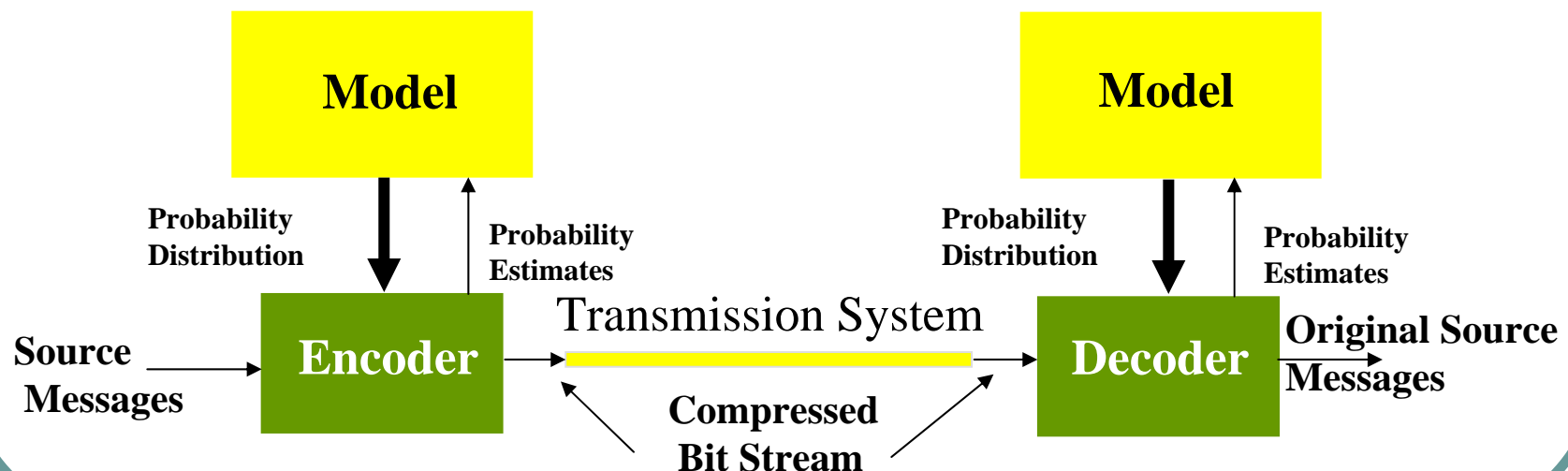
- Introduction to Image Compression
- Transform Coding
- **Subband Coding, Filter Banks**
- Introduction to Wavelet Transform
- Haar, SPIHT, EZW, JPEG 2000
- Motion Compensation
- Wireless Video Compression

# Contents

- From DCT/DFT to Subband Coding
- Filters
  - Low pass filters
  - High pass filters
  - Filter banks
  - Ideal filters
  - Realistic filters
  - Impulse Response of the Filters
  - Quadrature Mirror Filters
- Application of Subband Coding
- Relationship to Wavelet Transform Coding

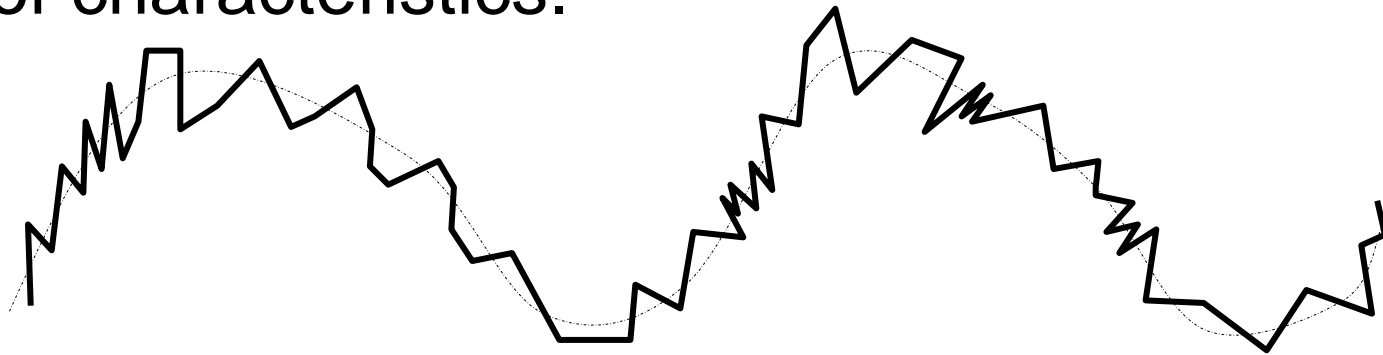
# Model and Prediction

- Compression is **PREDICTION**.
- Good model results in better prediction.
  - For text, we have good models (PPM, BWT).
  - For images, we are in need of a good model!



# Separate the Signals

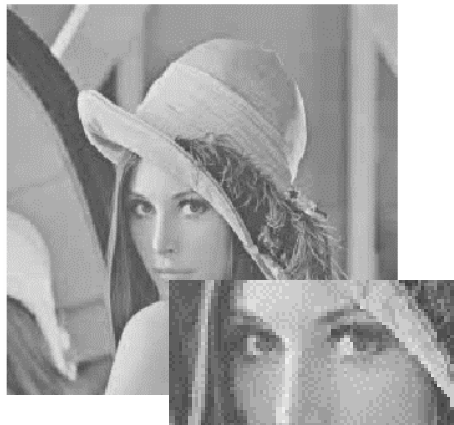
- Modeling a general signal is difficult.
- However, most signals exhibit a combination of characteristics.



- This signal can be decomposed into two signals (long term and short term), each can be simulated by an appropriate model.

# DCT/DFT Spatially Decomposition

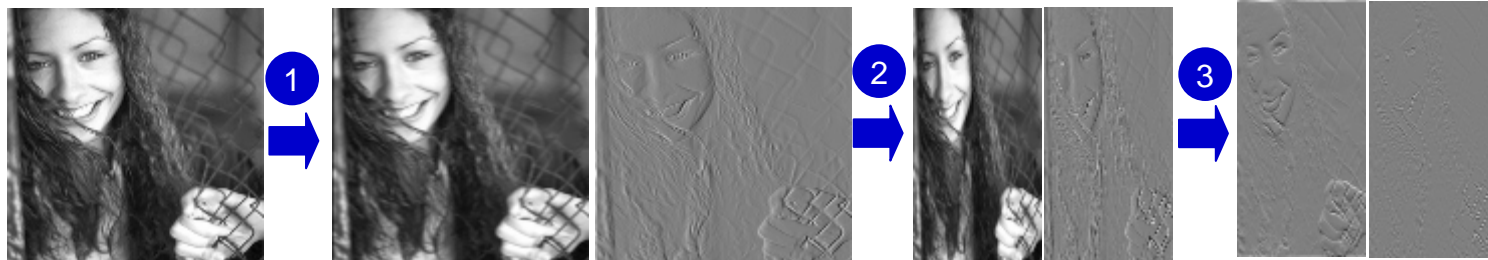
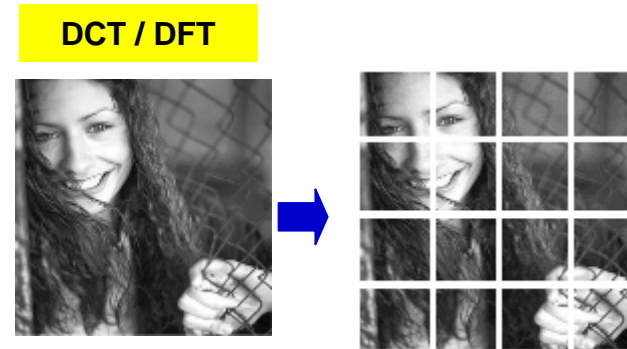
- DCT/DFT spatially decompose the image into blocks.
  - Perform frequency domain processing in each block.
  - Discard high frequency information in each block, because human eyes are insensitive to high frequencies.
  - However, there are artifacts at the block edges.



Coarser  
quantization

# Frequency Domain Decomposition

- Instead of spatially divide the image into blocks (such as DCT/DFT), we can consider frequency domain processing on the whole image.
  - Filtering the image horizontally into two subbands (images). ①
  - Down-sampling horizontally. ②
  - Filtering subbands vertically. ③
  - Down-sampling vertically. ④
  - Model and code each subbands separately.
- This is the subband coding.



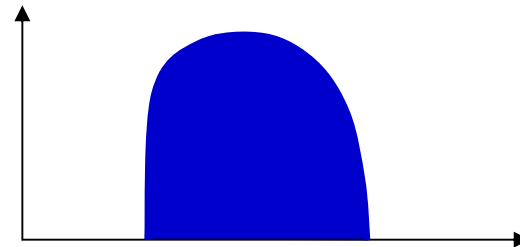
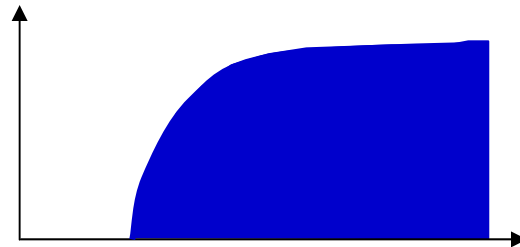
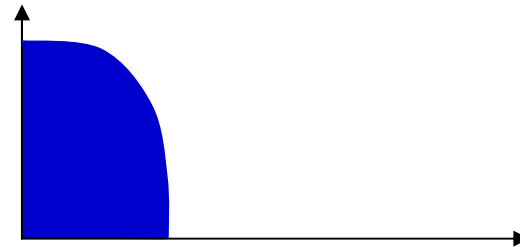
# Subband Coding

- Subband coding is a technique of decomposing the source signal into constituent parts and decoding the parts separately.
- A system that isolates a constituent part corresponding to certain frequency is called a *filter*. If it isolates the low frequency components, it is called a *low-pass filter*. Similarly, we have *high-pass or band-pass filters*. In general, a filter can be called a *sub-band filter* if it isolates a number of bands simultaneously.



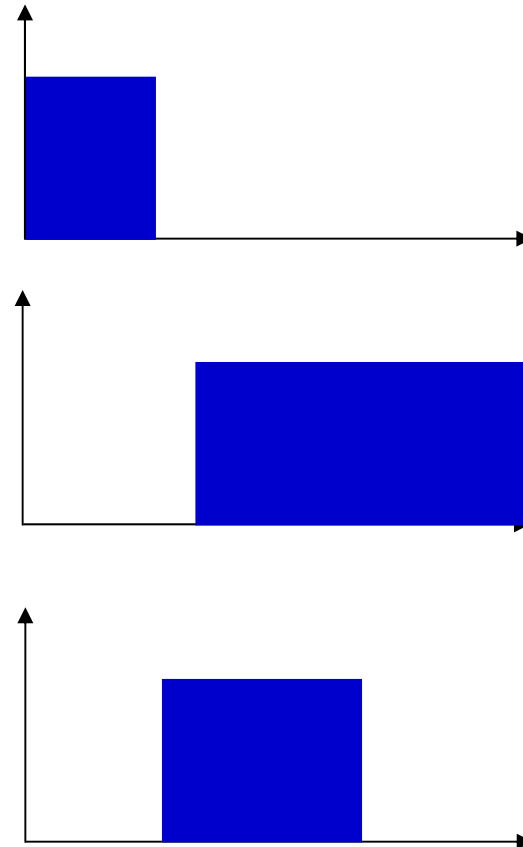
# Filters

- A system that isolates certain frequency components is called a filter.
  - Low-pass filter
    - Only let through components below a certain frequency  $f_0$
  - High-pass filter
    - Block all frequency components below a certain frequency  $f_0$
  - Band-pass filter
    - Let through  $[f_1, f_2]$



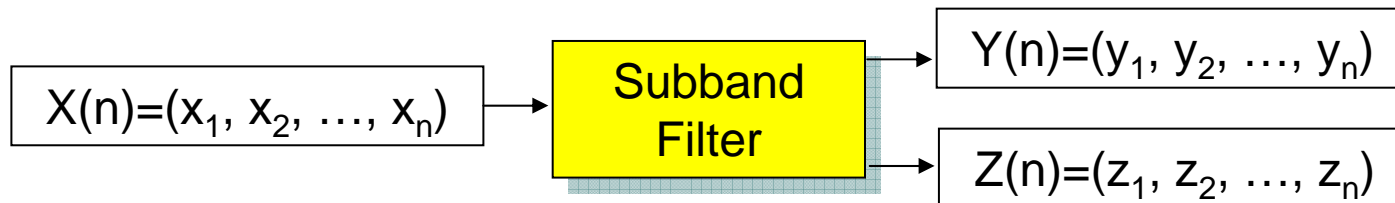
# Ideal Filters

- Ideal filters have a clear cut-off frequency  $f_0$ .
  - Ideal low-pass filter
  - Ideal high-pass filter
  - Ideal band-pass filter
- Ideal filters are unrealizable in Electrical Engineering.
- However, we have “perfect” Ideal Filters in computer science.
  - Can be implemented in any programming language.



# Subband Coding

- Consider the following scheme for a simplified signal system:



where

$$y_1 = x_1, y_2 = (x_2 + x_1)/2, \dots, y_n = (x_n + x_{n-1})/2$$

$$z_1 = x_1 - y_1, z_2 = x_2 - y_2, \dots, z_n = x_n - y_n$$

# Subband Coding

- In general, we can write

$$y_i = \frac{x_i + x_{i-1}}{2} \quad (\text{average})$$

$$z_i = \frac{x_i - x_{i-1}}{2} \quad (\text{difference})$$

for  $1 \leq i \leq n$  and  $x_0 = 0$ .

- The original signal can be recovered as

$$x_i = y_i + z_i$$

with  $y_1 = x_1$  and  $z_1 = 0$ .

# Subband Coding

- The signals  $y_n$  being averages, are much more smooth (lower frequency) and if the signals are correlated DPCM will be very effective.
- Similarly, the dynamic range of variation of  $z_n$  will be small. In fact, it is possible for the same number of bits per sample to encode both  $y_i$  and  $z_i$  with less distortion.

# Subband Coding

- But, there is a problem in the above scheme ---- we now have to use vectors  $Y(n)$  and  $Z(n)$ , each having  $n$  values ---- we have doubled the number of output elements!
- Let's divide  $\{y_i\}$  into two sets  $\{y_{2i}\}, \{y_{2i-1}\}$ :  
 $\{y_{2i-1}\} = \{y_1, y_3, y_5, \dots\}$  -- odd sequence  
 $\{y_{2i}\} = \{y_2, y_4, y_6, \dots\}$  -- even sequence

Similarly,

$$\begin{aligned}\{z_{2i-1}\} &= \{z_1, z_3, z_5, \dots\} && \text{-- odd sequence} \\ \{z_{2i}\} &= \{z_2, z_4, z_6, \dots\} && \text{-- even sequence}\end{aligned}$$

and transmit only the odd or even numbered sequence only.

# Subband Coding

- Suppose we only transmit the even sequence, we know

$$y_{2i} = \frac{x_{2i} + x_{2i-1}}{2} \quad z_{2i} = \frac{x_{2i} - x_{2i-1}}{2}$$

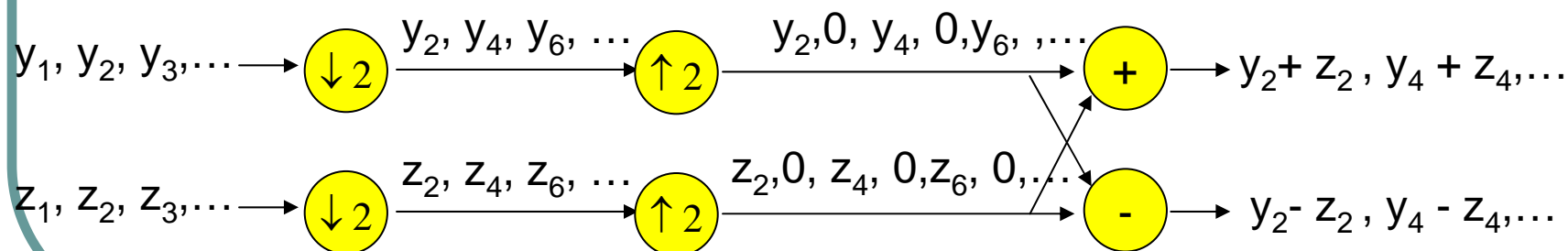
- Then

$$x_{2i} = y_{2i} + z_{2i} \quad x_{2i-1} = y_{2i} - z_{2i}$$

- Both odd ( $x_{2i-1}$ ) and even ( $x_{2i}$ ) numbered sequences are recovered. We can again use DPCM to transmit the odd or even signal attaining effective encoding as before.

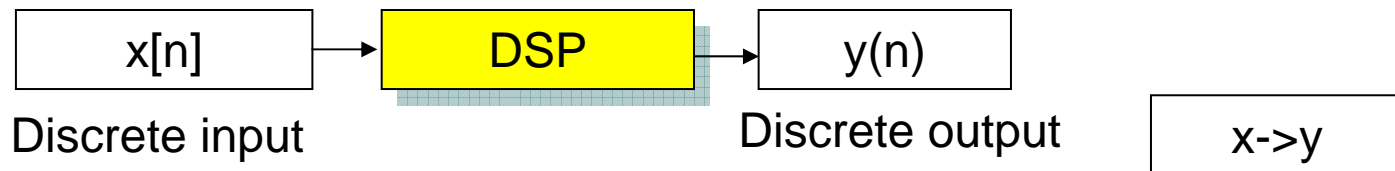
# Down-sampling and Up-sampling

- The process of transmitting either the odd or even sequence is called down-sampling, or decimation, denoted as  $\downarrow 2$
- The original sequence can be recovered from the two down-sampled sequences by inserting 0's between consecutive samples of the two sequences, delaying one set of signals by one sample and adding two signals correspondingly. The process is called up-sampling, denoted as  $\uparrow 2$





# Quick Tutorial on DSP



$x(t)$ : continuous time signal for all  $t$ , positive or negative.

Discrete time signal:

$$x[n] = x(nT), n = 0, 1, 2, \dots, -1, -2, \dots$$

When  $T$  is the sampling period,

$$f_s = 1/T = \text{sampling frequency}$$

# Nyquist Theorem

- To faithfully reconstruct  $x(t)$  from  $x[n]$ , we must have  $f_s \geq 2f_{\max}$ , where  $f_{\max}$  is the maximum frequency in the signal  $x(t)$ .
- Sampling a signal at a frequency less than  $2f_{\max}$  might introduce obvious low frequency aliased signals at the output.
- To satisfy this, it is a common practice to use a low pass filter to purge the signal of all frequencies greater than half of the maximum sampling frequency.

# Operation on Signals: Linear System

- Delay:  $x[n-k]$  ( $x[n]$  delayed by  $k$  sample periods)
- Add constant:  $x[n] + c$
- Multiply by a scalar:  $a x[n]$
- Summation :  $x_1[n] + x_2[n]$

# Linear System

- Superposition:  
if  $x_1 \rightarrow y_1$ ,  $x_2 \rightarrow y_2$ , then  $x_1 + x_2 \rightarrow y_1 + y_2$
- Homogeneity:  
if  $x \rightarrow y$ , then  $cx \rightarrow cy$

# Realistic Filtering

- Output  $y[n]$  is calculated by taking a weighted sum of the following into the filter
  - current input  $x[n]$
  - past inputs ( $x[0], x[1], \dots, x[n-1]$ )
  - in some cases, the past outputs ( $y[0], y[1], \dots, y[n-1]$ ) of the filter.
  - $a_i, b_i$  are *filtering coefficients*.

$$y[n] = \sum_{i=0}^N a_i x[n-i] + \sum_{i=1}^M b_i y[n-i]$$

# Impulse Response of the Filter

- If the input sequence is a single 1 followed by all 0s (1000...0), the output sequence is called the impulse response of the filter (represented as  $\{h_n\}$ ). The input sequence is called **Impulse (delta) function**.

- Finite impulse response filter (FIR)

- All  $b_i$  are 0.
- The impulse will die out after N samples.
- N is the number of taps in the filter.

$$y[n] = \sum_{i=0}^N a_i x[n-i]$$

- Infinite impulse response filter (IIR)

- Any of the  $b_i$  is not 0.
- The impulse can continue forever.

$$y[n] = \sum_{i=0}^N a_i x[n-i] + \sum_{i=1}^M b_i y[n-i]$$

# Calculate Impulse Response

- Assume  $a_0=1$ ,  $b_1 = 0.9$ , all other  $a_i$ ,  $b_i$  are 0.
- Input  $x[0] = 1$ ,  $x[1] = x[2] = x[3] = \dots = x[n] = 0$

- Output:

- $y_0 = a_0 x[0] = 1 * 1 = 1$
- $y_1 = a_0 x[1] + b_1 y[0] = 1 * 0 + 0.9 * 1 = 0.9$
- $y_2 = a_0 x[2] + b_1 y[1] = 1 * 0 + 0.9 * 0.9 = 0.81$
- ...
- $y_n = a_0 x[n] + b_1 y[n-1] = 1 * 0 + 0.9 * 0.9^{n-1} = 0.9^n$

$$y[n] = \sum_{i=0}^N a_i x[n-i] + \sum_{i=1}^M b_i y[n-i]$$

- So, we have the impulse response

$$h_n = \begin{cases} 0 & n < 0 \\ 0.9^n & n \geq 0 \end{cases}$$

# Restore the Signal from the Impulse Response Function

- Output  $\{y_n\}$  can be restored from the input  $\{x_n\}$  and the impulse response function  $\{h_n\}$ :

$$y[n] = \sum_{k=0}^M h_k x[n-k]$$

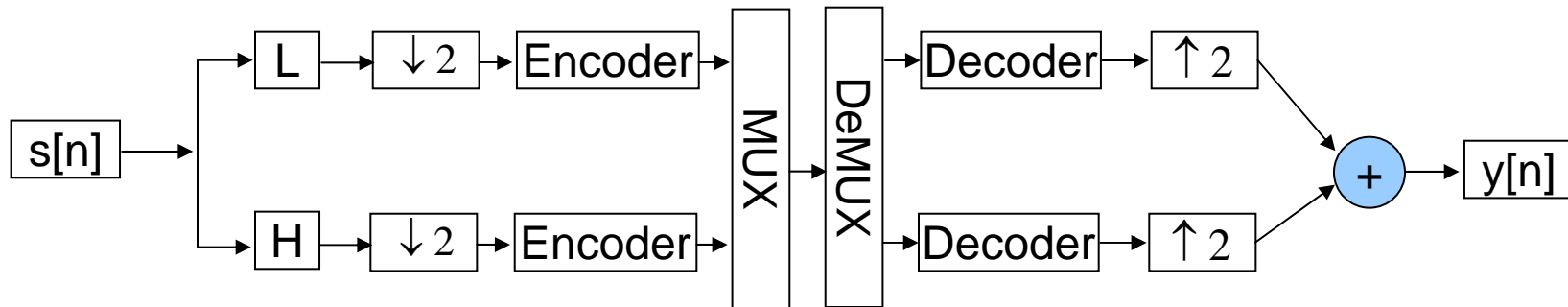
- M is
  - finite for the FIR
  - Infinite for the IIR
- Termed convolution.
  - Good for hardware implementation.



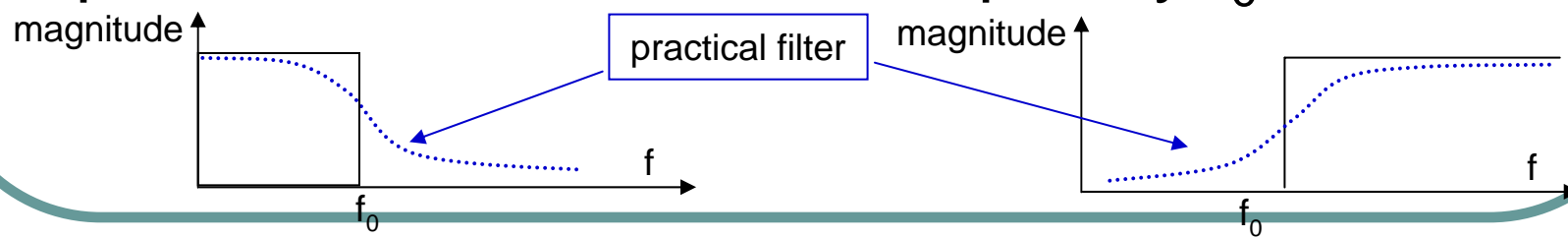
# Subband Coding

- The principle of splitting a discrete time signal into a number of subband signals and combining the subband signals into final output signal has led to development of 'filter bank' system of analysis and synthesis for discrete signal processing (DSP).
- The basic two channel subband codec (coder/decoder) based on a two-channel QMF (quadrature mirror filter) bank is shown below.

# Quadrature Mirror Filter



- The decomposition is usually done in the frequency domain. The input signal  $x[n]$  is first passed through a two-band analysis filter bank containing  $L$ , a low-pass filter and  $H$ , a high pass filter, with a cut-off frequency  $f_0$ .



# Quadrature Mirror Filter

- The subband signals are down-sampled (decimated) by a factor of 2. Each down-sampled subband signal is encoded/quantized by special characteristics of the signal, such as energy level and perceptual importance. The coded subband are then combined into one sequence by a multiplexer.
- At the receiving end, the coded subband signals are first recovered by demultiplexing.

# Quadrature Mirror Filters

- The subband coding uses a cascade of stages.
- Each stage contains a pair of lowpass and highpass filters. The most commonly used filters are called Quadrature Mirror Filter (QMF).
- QMF Filters are mirror symmetric.
  - Impulse response of the low pass filter  $\{h_n\}$
  - Impulse response of the high pass filter  $\{H_n\}$

$$H_n = [-1]^n h_{N-1-n} \quad n = 0, 1, \dots, N-1$$

$$h_{N-1-n} = h_n \quad n = 0, 1, \dots, \frac{N}{2} - 1$$

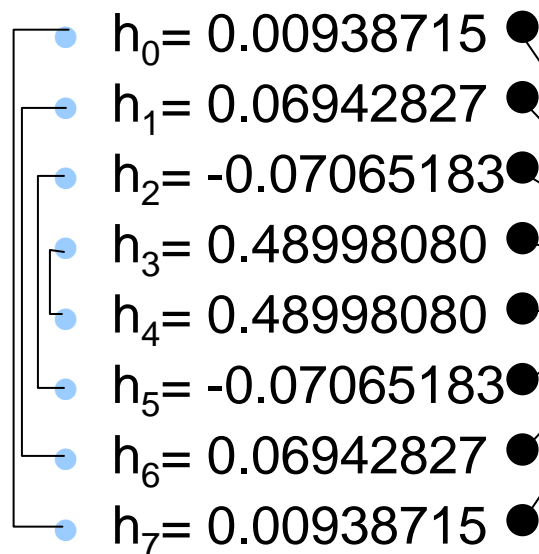
# Example of Quadrature Mirror Filters

$$H_n = [-1]^n h_{N-1-n} \quad n = 0, 1, \dots, N-1$$

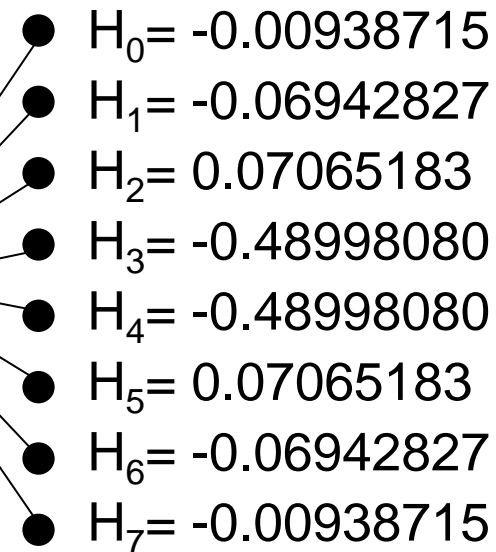
$$h_{N-1-n} = h_n \quad n = 0, 1, \dots, \frac{N}{2} - 1$$

- Tap  $N = 8$ .

- Low pass



- High pass



# Subband Coding

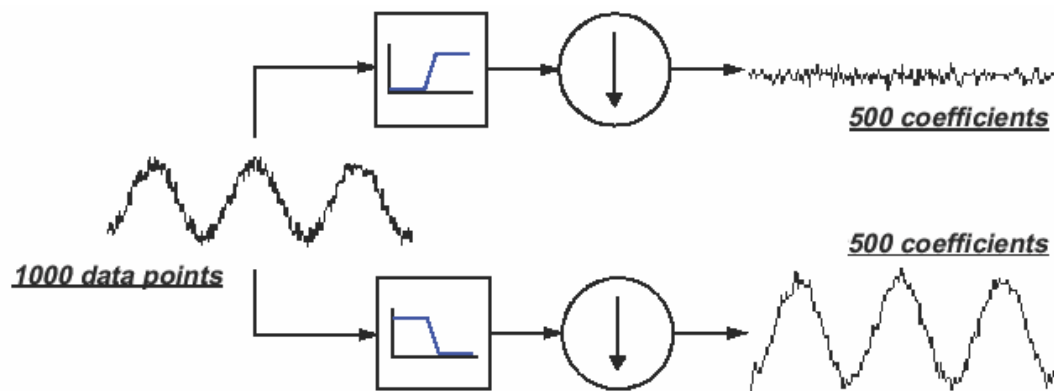
- Image is first filtered to create a set of images:
  - Each image contains a limited range of spatial frequencies.
  - Images are called subbands.
  - Each subband has a reduced bandwidth compared to the original full-band image.
  - The image may be down-sampled.
  - Filtering and sub-sampling is termed analysis stage.
- Each subband is encoded
  - using one coder.
  - or, using multiple coders (recursive subband coding).

# Subband Coding

- Different bit rates or even different coding techniques can be used for each subband.
  - Takes advantages of properties of the subband.
  - Allowing for the coding errors to be distributed across the subbands in a visually optimal manner.
    - Error in one subband will be distributed onto the whole image in the reconstruction stage.
    - In construction phase, error in the DCT is confined in each blocks.
- Reconstruction is achieved by up-sampling the decoded subbands, applying appropriate filters and adding the reconstructed subbands together
  - Termed synthesis stage

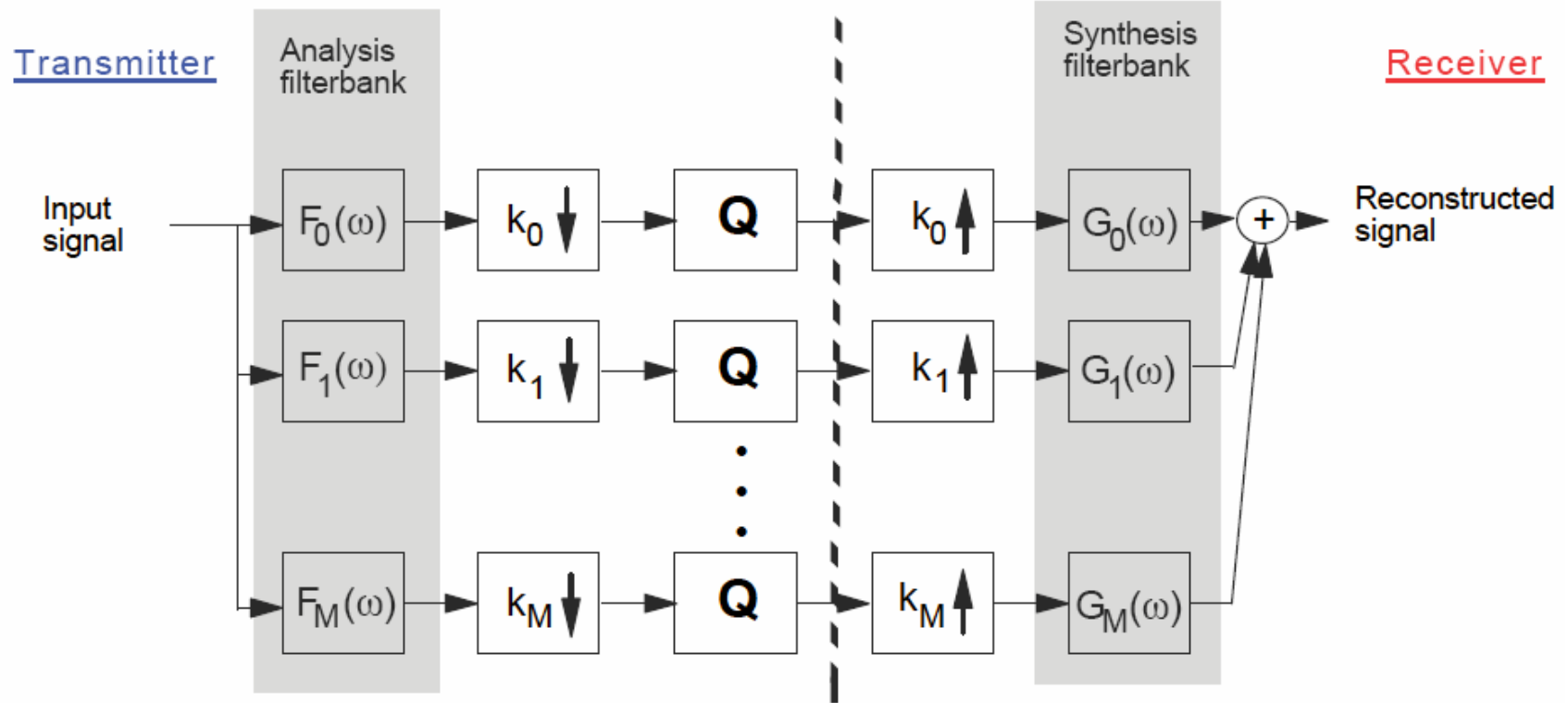
# Subband Coding

- Formulation of subbands **DOES NOT** create any compression.
  - Same number of samples is required to represent the subbands as is required for the original images.
- Subbands can be encoded more efficiently than the original image.

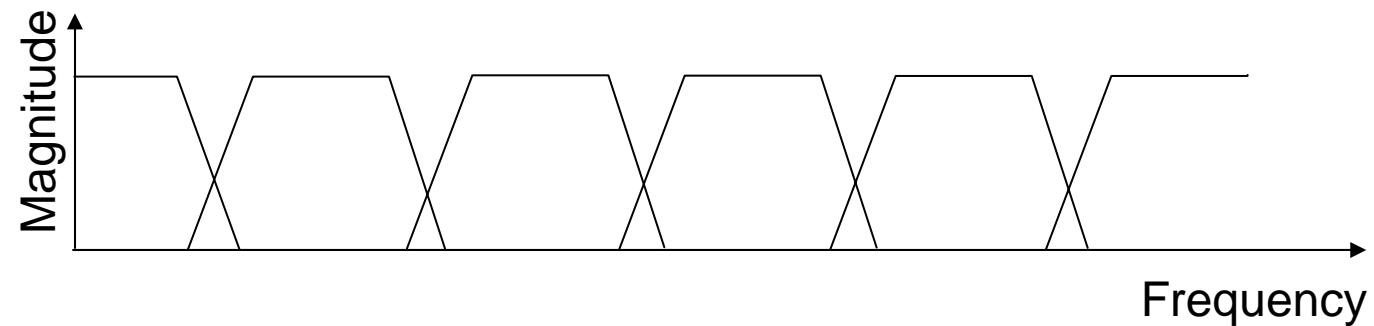
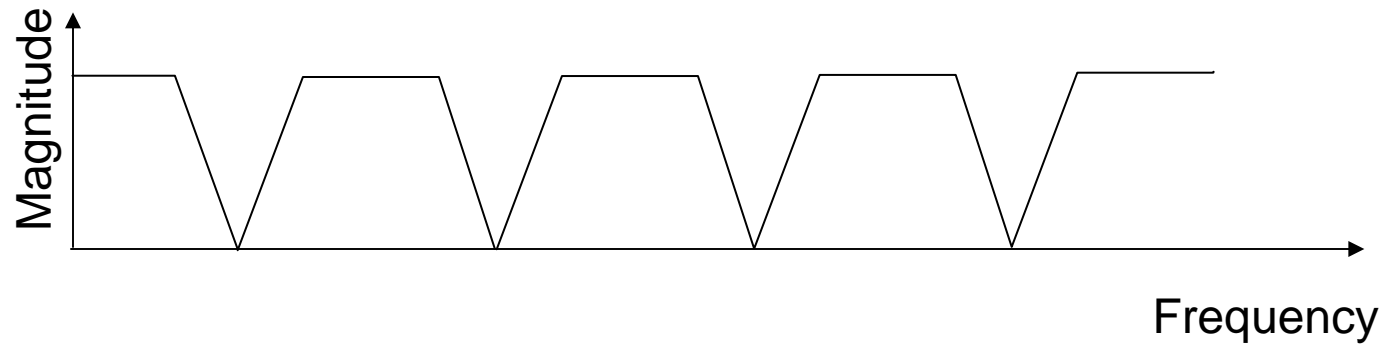




# 1D Multi-band Coding

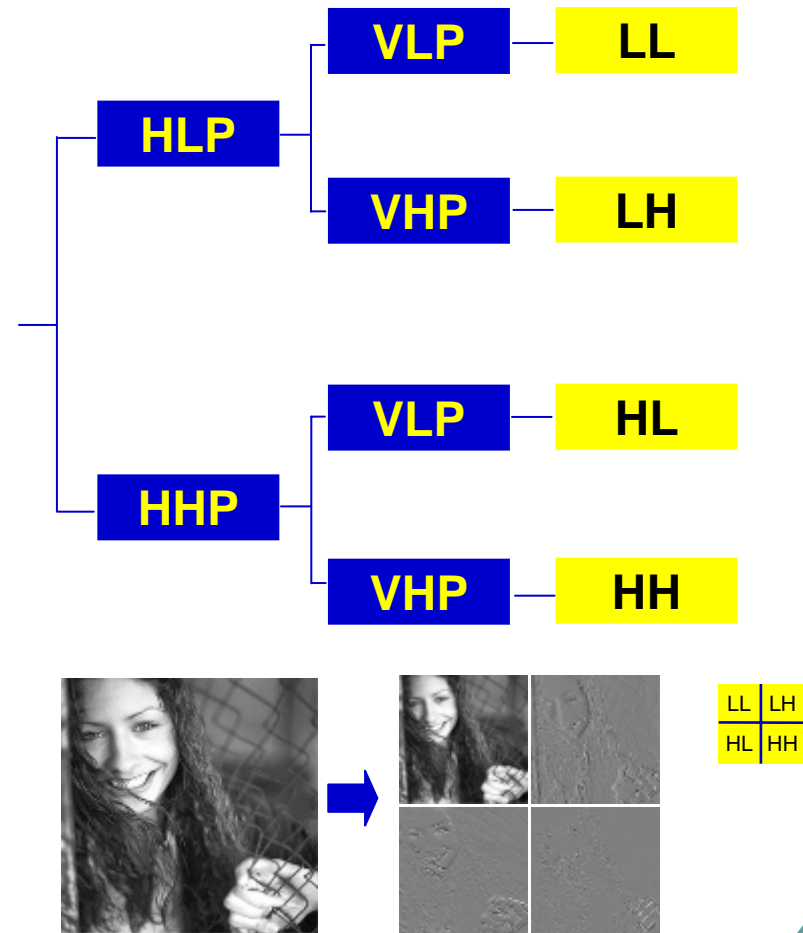


# Nonoverlapping/overlapping Filter Banks



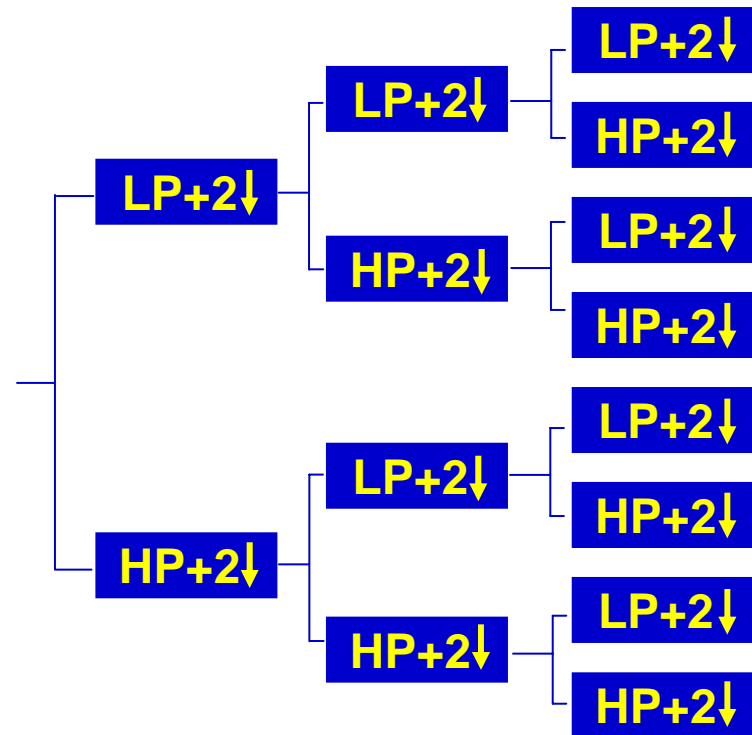
# 2D, Two-band Subband Coding

- Separable filter
  - Horizontal low-pass filter (HLP)
  - Horizontal high-pass filter (HHP)
  - Vertical low-pass filter (VLP)
  - Vertical high-pass filter (VHP)
- YES. There are 2D nonseparable filters.
  - Drawback: high computational complexity.



# Cascaded Filter Banks

- The most frequently used filter banks in subband coding consist of a cascade of images
  - Each stage consists of a low-pass filter and a high-pass filter.
  - Down sampling.



# Haar Transform and Filter Banks

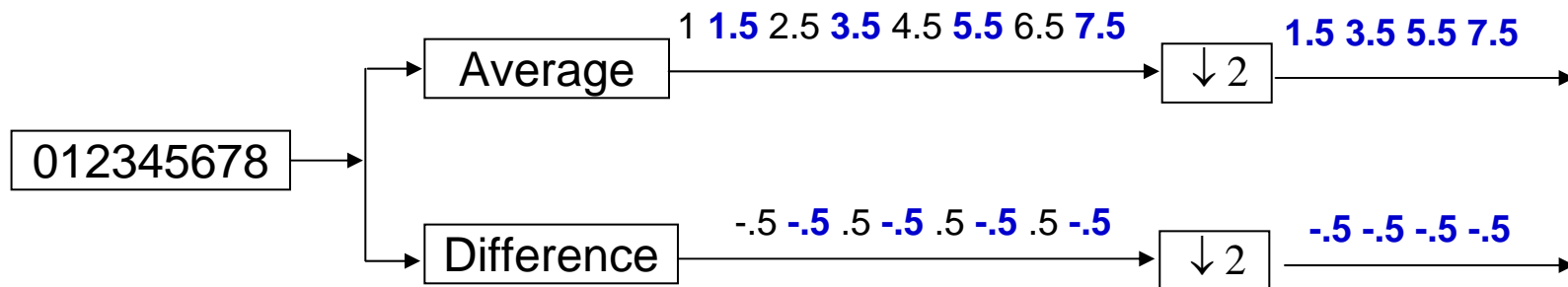
- We have seen the convolution operator earlier

$$y[n] = \sum_{k=0}^M h_k x[n-k]$$

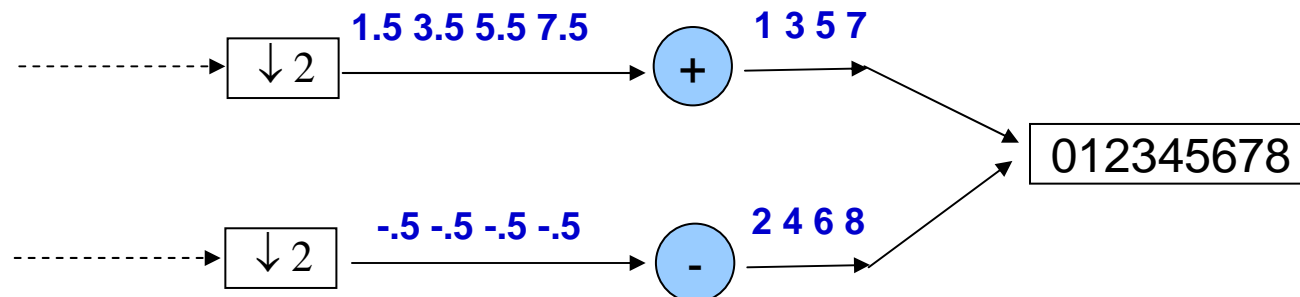
where  $h_0, h_1, \dots$  are the filter coefficients.

- Let's define two 2-tap filters as follows
- L:  $y[n] = 0.5 * x[n] + 0.5 * x[n-1]$   
H:  $y[n] = 0.5 * x[n] - 0.5 * x[n-1]$
- L is the 'average' of signal value  $x[n]$  with the one unit delayed  $x[n-1]$  producing  $y[n]$ . L is recognized as a 'low-pass' filter. Similarly, H is the 'difference' function recognized as a 'high-pass' filter.

# Down-sampling and Up-sampling

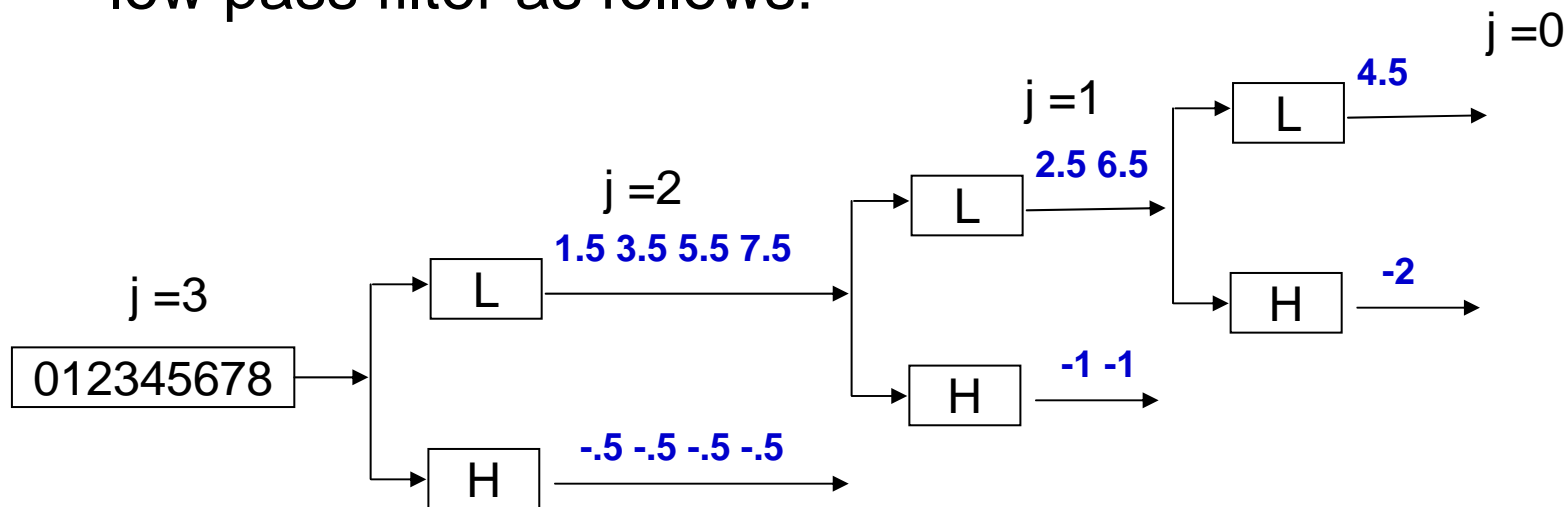


- We really do not need to compute the decimated output.
- Up-sampling and Synthesis Filter



# Logarithmic Tree

- We can iterate the filter operations on the output of the low pass filter as follows:



- Output: 4.5 -2 -1 -1 -.5 -.5 -.5 -.5
- The logarithmic tree is also called the multiresolution tree.

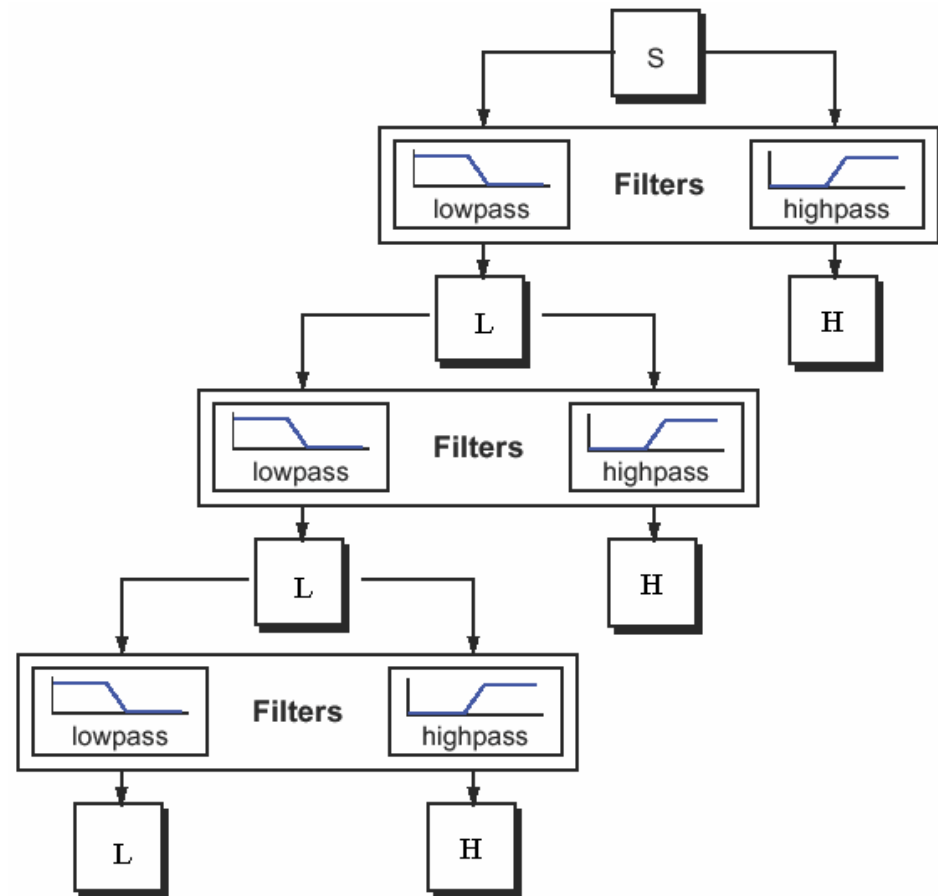
# One-dimensional Haar Wavelet Transform

- Given input value {1, 2, 3, 4, 5, 6, 7, 8} (resolution 8)
- Step #1 (resolution 4)
  - Output Low Frequency {1.5, 3.5, 5.5, 7.5} - average
  - Output High Frequency {-0.5, -0.5, -0.5, -0.5} – detail coefficients
- Step #2 (resolution 2)
  - Refine Low frequency output in Step #1
    - L: {2.5, 6.5}- average
    - H: {-1, -1} - detail
- Step #3 (resolution 1)
  - Refine Low frequency output in Step #2
    - L: {4.5} -average
    - H: {-2} - detail
    - Transmit { 4.5, -2, -1, -1, -0.5, -0.5, -0.5, -0.5}. No information has been lost or gained by this process. We can reconstruct the original image from this vector by adding and subtracting the detail coefficients. The vector has 8 values, as in the original sequence, but except for the first coefficient, all have small magnitudes..



# Sub-band Interpretation of Wavelet Transform

- The computation of the wavelet transform used recursive averaging and differencing coefficients. It behaves like a **filter bank**.
- Recursive application of a two-band filter bank to the lowpass band of the previous stage.



# Matrix Operation

- The discrete wavelet transform can also be described in terms of matrix operation.
- The Haar filter operation for  $j = 1$  is equivalent to multiplying the two input by the matrix

$$\frac{1}{\sqrt{2}} \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$$

# Matrix Operation

- Similarly, for  $j = 2$  and  $j = 3$ , the corresponding matrices are

$$\frac{1}{\sqrt{4}} \begin{bmatrix} 1 & 1 & 0 & 0 \\ 0 & 0 & 1 & 1 \\ 1 & -1 & 0 & 0 \\ 0 & 0 & 1 & -1 \end{bmatrix}$$

$$\frac{1}{\sqrt{8}} \begin{bmatrix} 1 & 1 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 \\ 1 & -1 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & -1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & -1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & -1 \end{bmatrix}$$

# Generalization of other filters

- The matrix formulation extends to other kinds of filters as well. For example, the Daubechi's D4 filter has four coefficients:

$$c_0 = 0.48296$$

$$c_1 = 0.8365$$

$$c_2 = 0.2241$$

$$c_3 = -0.1294$$

- The transform matrix has the form:

$$w = \begin{bmatrix} c_0 & c_1 & c_2 & c_3 & 0 & 0 & \dots & 0 \\ c_3 & -c_2 & c_1 & -c_0 & 0 & 0 & \dots & 0 \\ 0 & 0 & c_0 & c_1 & c_2 & c_3 & \dots & 0 \\ 0 & 0 & c_3 & -c_2 & c_1 & -c_0 & \dots & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \dots & 0 & c_0 & c_1 & c_2 & c_3 \\ 0 & 0 & \dots & 0 & c_3 & -c_2 & c_1 & -c_0 \\ c_2 & c_3 & 0 & 0 & \dots & 0 & c_0 & c_1 \\ c_1 & -c_0 & 0 & 0 & \dots & 0 & c_3 & -c_2 \end{bmatrix}$$

# Generalization of other filters

- If the input vector  $X = (x_1, x_2, \dots, x_n)$  is multiplied with  $W$ , we get

- H: smooth coefficients

$$s_1 = c_0x_1 + c_1x_2 + c_2x_3 + c_3x_4$$

$$s_3 = c_0x_3 + c_1x_4 + c_2x_5 + c_3x_6, \text{ etc.}$$

- G: detail coefficients

$$d_1 = c_3x_1 - c_2x_2 + c_1x_3 - c_0x_4$$

$$d_3 = c_3x_3 - c_2x_4 + c_1x_5 - c_0x_6, \text{ etc.}$$

# Generalization of other filters

- Note, these smooth and detail coefficients are convolutions of data with the four coefficients. Together  $H$  and  $G$  form a QMF.
- The coefficient values have been derived from orthonormality conditions and therefore, the inverse of  $W$  is  $W^T$ .
- Since the size of  $W$  is that of the image, it may seem impractical from storage point of view. But, the matrix is very regular. Given the top row of  $W$ , all other rows can be generated by simple shifting, reversing and changing signs.

# Coding of the Subbands

- Coding of the subbands is done using a method and bit rate most suitable to the statistics and visual significance of that subband.
- Typically 95% of the image energy is in the low frequency bands.
- Low bands may use Transform, DPCM or VQ.
- Other bands may use PCM or run-length coded after coarse thresholding.

# Application of Subband Coding

- Speech coding
  - ITU-T G.722
  - Encode high-quality speech at 64/56/48 kbps.
- Audio coding
  - MPEG Audio
    - Layer 1
    - Layer 2
    - Layer 3
- Image compression
  - Closely related to wavelet transform coding.



# Subband Coding and Wavelet Transform Coding

- Filters used in subband coders are not in general orthogonal.
- Transform coding is a special case of subband coding
- Wavelet transform coding is closely related to subband coding.
  - Alternative to filter design.
- Wavelet transform describes multiresolution decomposition in terms of expansion of an image onto a set of wavelet basis functions.
  - Basis functions well localized in both space and time.
- Wavelet transform-based filters possess some regularity properties that not all QMF filters have.
  - Improved coding performance over QMF filters with same number of taps.