

Fourier Transform

Decomposition of signals in terms of sinusoids

$$A^{i\omega x} = A \cos(\omega x) + iA \sin(\omega x)$$

Using complex exponential as an input

$$g[x] = e^{i\omega x} * h[x] = \sum_{k=-\infty}^{\infty} h[k] e^{i\omega(x-k)} = e^{i\omega x} \sum_{k=-\infty}^{\infty} h[k] e^{-i\omega k}$$

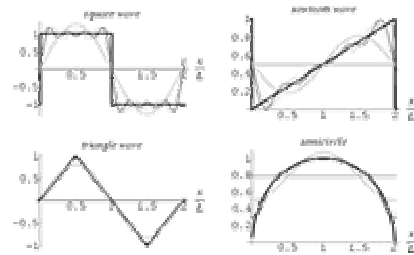
$$g[x] = e^{i\omega x} H(\omega)$$

Given complex exponential as input, output is again complex exponential scaled by $H(\omega)$

$H(\omega)$ is the frequency response of linear-time Invariant systems

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Fourier Series



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Fourier Series and Transform

Decomposition of signals in terms of sinusoids - Fourier series

$$f[x] = \frac{1}{2\pi} \sum_{j=-\infty}^{\infty} c_j \cos[kx + \phi_k]$$

$$f[x] = \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} a_k \cos[kx] + b_k \sin[kx] \quad a_k = \sum_{j=-\infty}^{\infty} f[j] \cos[kj] \quad b_k = \sum_{j=-\infty}^{\infty} f[j] \sin[kj]$$

Fourier series and transform

$$f[x] = \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} c_k e^{ikx} \quad \text{and} \quad c_k = \sum_{j=-\infty}^{\infty} f[j] e^{-ij k}$$

Alternative form

$$f[x] = \frac{1}{2\pi} \sum_{\omega=-\pi}^{\pi} F[\omega] e^{i\omega x} \quad \text{and} \quad F[\omega] = \sum_{k=-\infty}^{\infty} f[k] e^{-i\omega k}$$

Central theme - approximate a function with given a family of basis functions

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Fourier Transform

$$F[\omega] = \sum_{k=-\infty}^{\infty} f[k] e^{-i\omega k} \quad |F[\omega]| = \sqrt{F_i[\omega]^2 + F_r[\omega]^2} \\ \angle F[\omega] = \tan^{-1}(F_i(\omega)/F_r(\omega))$$

- Connecting back to LTI
- Frequency response of a linear time invariant system is the Fourier transform of the unit-impulse response
- i.e. LTI are uniquely characterized by their impulse response and equivalently by their frequency response
- Relationship between frequency and space domains

$$g[x] = f[x] * h[x] \quad G[\omega] = F[\omega]H[\omega]$$

$$F = Mf$$

$$M = \begin{bmatrix} 1 & 1 & 1 & \dots & 1 \\ e^{-0i} & e^i & e^{-2i} & \dots & e^{-7i} \\ e^{-0i} & e^{2i} & e^{-4i} & \dots & e^{-27i} \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ e^{-0i} & e^{7i} & e^{-27i} & \dots & e^{-7^2i} \end{bmatrix}$$

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Fourier transform as representation

- The spatial information is lost - spectral information is gained
- Can we achieve simultaneous localization in space (time) and frequency? -> only within some bounds
- Windowed Fourier Transform - balance between the two

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Filtering - Fourier Transform

- Examples FT phase and frequency information

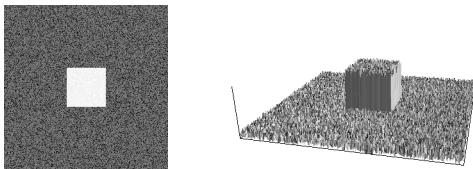
$$\begin{array}{ccc} \xrightarrow{f} & \boxed{\text{filter}} & \xrightarrow{g} \\ g[x] = f[x] * h[x] & & G[\omega] = F[\omega]H[\omega] \end{array}$$

- Low pass filtering
- High pass filtering
- Band pass filtering
- Differentiation

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Fourier transform and filtering

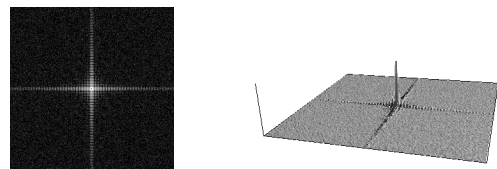
- <http://astronomy.swin.edu.au/~pbourke/analysis/imagefilter/>



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Fourier transform and denoising

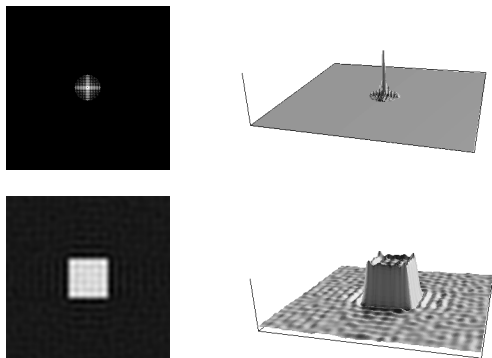
- <http://astronomy.swin.edu.au/~pbourke/analysis/imagefilter/>



Fourier transform of the image

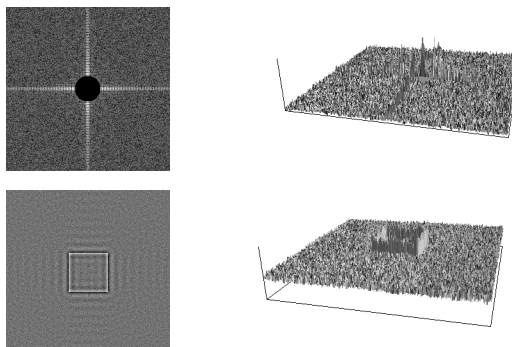
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Low Pass filtering



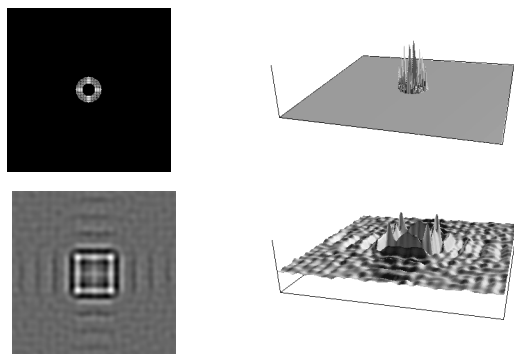
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High pass filtering



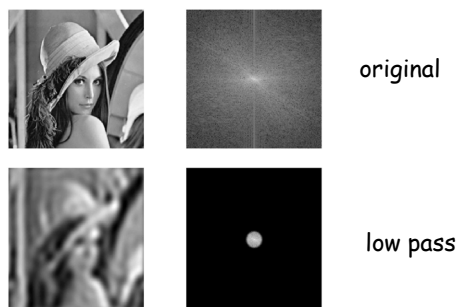
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Band-pass filtering



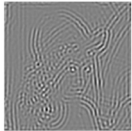
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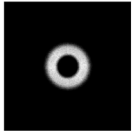
FFT by example




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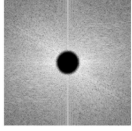
FFT Example





bandpass

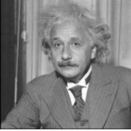




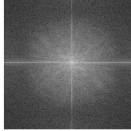
highpass

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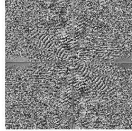
Importance of phase




original



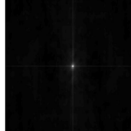
power



phase



constant power + original phase



original power + zero phase

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Scale Space Representations

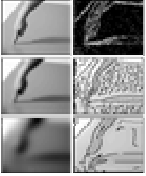
- **Gaussian Pyramids**
enable to extract different structures in the image - since different structures are more apparent at different scales (useful for search over scale - detection, spatial search, feature tracking)

$$I(x, y, \sigma) = I_0(x, y) * G(x, y, \sigma)$$

(Witkin'83, Koenderink, van Doorn'86)

Family of functions - solutions to diffusion eq.

$$\frac{\partial I}{\partial \sigma} = \frac{\partial I}{\partial x^2} + \frac{\partial I}{\partial y^2} = \nabla^2 I$$



Courtesy T. Lindenberg, KTH

- Causality (features at coarse level - have cause at fine level)
- Homogeneity and Isotropy (blurring is space-invariant)

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Anisotropic Diffusion

Problem with traditional scale space models
Gaussian is symmetric - smoothes over edges
Does not preserve localization of edges
Idea - edge preserving smoothing

$$\frac{\partial I}{\partial \sigma} = \frac{\partial I}{\partial x^2} + \frac{\partial I}{\partial y^2} = \nabla^2 I \quad I(x, y, 0)$$

With spatially varying term (Perona & Malik '90)

$$\frac{\partial I}{\partial \sigma} = \nabla \cdot (c(x, y, \sigma) \nabla I) = c(x, y, \sigma) \nabla^2 I + (\nabla c(x, y, \sigma)) \cdot (\nabla I)$$

If we knew where are the edges - we can
Create a mask $c(x, y)$ (assuming that $c(x, y, \sigma)$ is independent of σ)

We get noise free image and smooth where there are no edges

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Anisotropic Diffusion - Example



Images courtesy P. Kovese (www.csee.unwa.au/~pk)

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